

Interference Reduction Based on Hyperbolic Fractional Fourier Transform in Integrated Sensing and Communication

Mohammad Reza Mousavi, Stephan Ludwig

Mohammad Reza Mousavi, Ph.D.

Mohammad.Mousavi@hs-aalen.de

Electrical Engineering Faculty

Aalen University of Applied Sciences

Germany

Contents

- The motivation
- Hyperbolic Fractional Fourier transform (HFrFT)
- Signal model and Analysis
- Simulation result
- Conclusion

The motivation:

- **Integrated sensing and communication (ISAC)**
 - The prominent application → toward 6G
 - Joint communication and sensing (JCAS)
 - Joint communication and radar (JCR)
 - Dual function radar communication (DFRC)
- Interference reduction → Common issue for telecommunication systems
 - Intended interference cancellation → security challenges of ISAC
- Fractional Fourier transform (FrFT) → Generalized form of Fourier transform → Chirp decomposition
 - FrFT-based Multicarrier systems
 - FrFT- based Radar signal processing
 - The main weakness → Only limited real and optimized orders of transform

Hyperbolic Fractional Fourier Transform (HFrFT)

- Continuous HFrFT:

$$\mathcal{F}_\alpha^h \{x(\tilde{t})\}(\tilde{v}) = \int_{-\infty}^{\infty} x(\tilde{t}) \cdot \overbrace{A \cdot e^{-\pi(\tilde{t}^2 + \tilde{v}^2) \coth(\alpha)}}^{\text{Transform kernel}} \cdot \underbrace{e^{2\pi\tilde{t}\tilde{v} \operatorname{csch}(\alpha)}}_{\text{wavelet shape}} d\tilde{t}$$

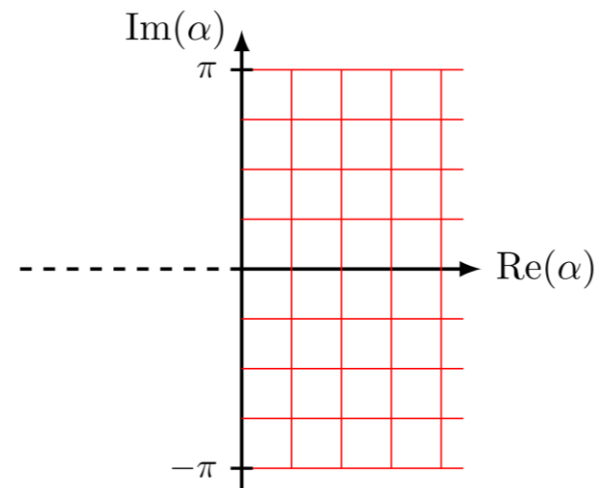
Gaussian shape

$$A = \frac{e^{-j(\frac{\pi}{4} - \frac{|\alpha|}{2})}}{\sqrt{2\pi j \sinh(\alpha)}}$$

$x(\tilde{t})$: The input signal

$\tilde{t} := t/s, \tilde{v} := v/\text{Hz}$

➤ For $\alpha = j \frac{\pi}{2}$ HFrFT \rightarrow Fourier transform (FT)



$$\alpha \in \mathbb{C} \left\{ \begin{array}{l} \Re(\alpha) > 0 \\ 0 < |\Im(\alpha)| < \pi \end{array} \right.$$

Signal model

$$r(t) = \sum_{l=0}^{L-1} h_l(t)y(t - lT) + \underbrace{\sum_{n=-\infty}^{\infty} \left(\frac{\tilde{A}\tau}{T_o} \right) \text{sinc}\left(\frac{n\tau}{T_o}\right) \cdot e^{j2n\pi \frac{t}{T_o}}}_{\text{Fourier series expansion of the Pulse train interference}} + \underbrace{\omega(t)}_{\text{AWGN noise}}$$

$$y(t) = \sum_{m=0}^{N-1} \hat{X}_\alpha^h[m]g(t - mT), \quad 0 \leq t < T_s$$

- \hat{X}_α^h : Transmitted symbol with subcarrier m
- g : baseband pulse shaping function
- T_s : Symbol interval
- h_l : l th path fading channel response
- \tilde{A} : Pulse amplitude
- τ : pulse width
- T_o : pulse period
- $\omega(t)$: AWGN noise

The received signal plus pulse train interference and AWGN noise

Simulation result

Transmitted

Interference

Received

OFDM

OFDM

HFrFT

HFrFT

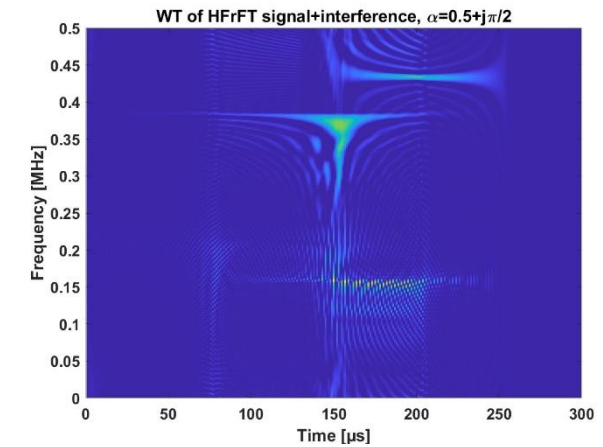
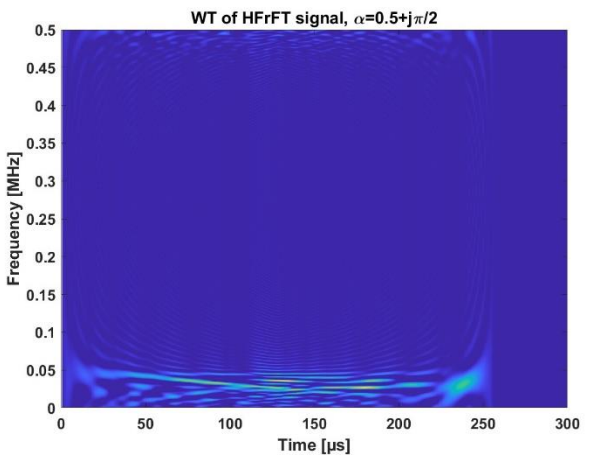
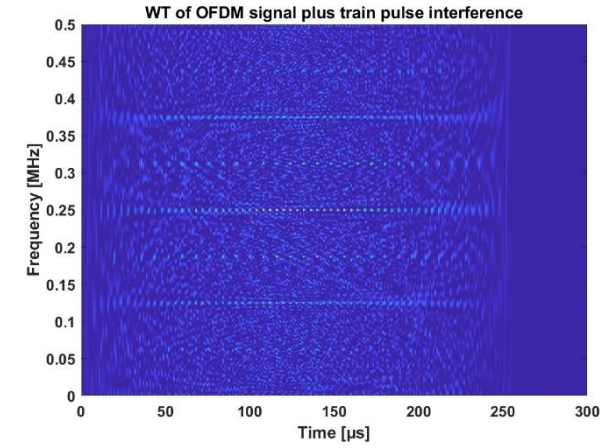
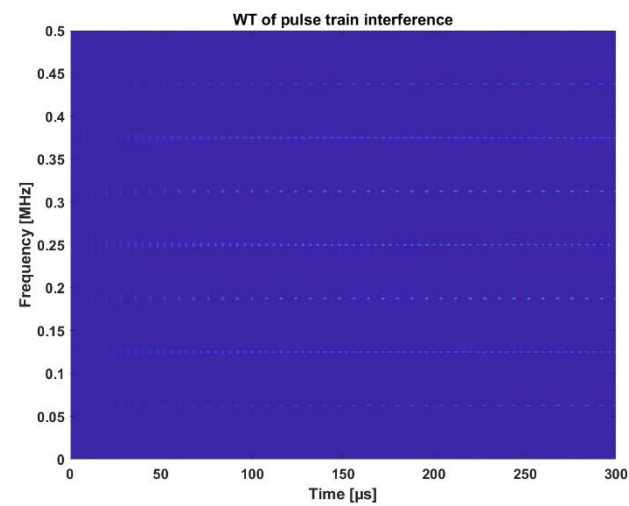
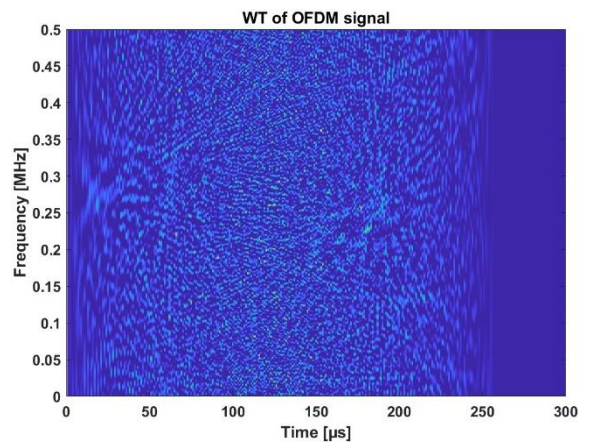


Fig. 1. Comparison between WT analysis of OFDM and HFrFT with the selected transform orders $\alpha = 0.5 + j\frac{\pi}{2}$ interfered with a pulse train interference

Simulation result

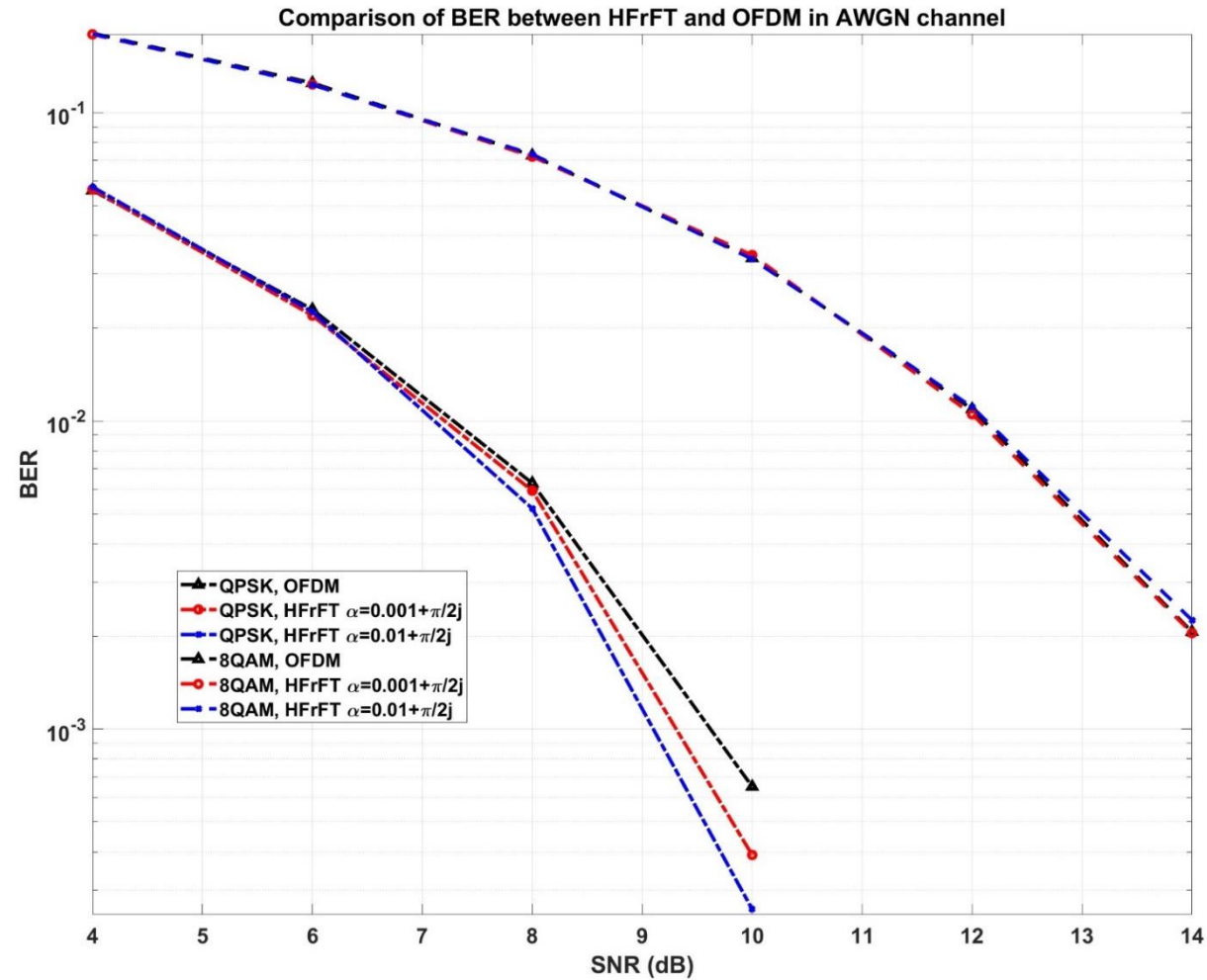


Fig. 2. Comparison of BER versus SNR for HFrFT and OFDM system in the AWGN channel for QPSK and QAM modulation schemes

Conclusion

- ❖ HFrFT results in a better BER performance especially in the QPSK scheme compared to ordinary OFDM
- ❖ Reducing the intended interference in HFrFT for the selected transform orders is easier compared to the ordinary OFDM
- ❖ Using the HFrFT technique in ISAC applications can produce better performance in both communication and sensing and introduce HFrFT as a suitable nomination in ISAC applications
- ❖ Open issues: Optimization of the transform order, the theoretical BER calculation for modulation schemes and different channels, and the throughput and complexity calculation for multi-user, multi-target, and MIMO scenarios

Thank You

Mohammad.Mousavi@hs-aalen.de